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### **Gravity Probe B Relativity Mission**

# **VERIFICATION OF T002 REQUIREMENT 5, Part A: LINEARITY OF GYROSCOPE READOUT SYSTEM**

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## **Table of Contents**

List of Tables	
List of Acronyms, Abbreviations, and Symbols	3
1. Introduction	4
2. Measurements of Harmonic Distortion	
3. Relationship of Nonlinearity to Harmonic Distortion	5
4. Effects of Nonlinearity on the Science Gyroscope Readout	<i>6</i>
4.1 Effects of Nonlinearity on the Signals at the Harmonics of the Gyroscope Spin Speed	6
5.2.1 Rectification of High Frequency Signals	6
5.2.2 Modulation of the Low Frequency Scale Factor	
4.2 Effects of Nonlinearity on Low Frequency Signals	7
5.3.1. Linearity of the Measurement Model	7
5.3.2 Changes in the Scale Factor That Depend on the Bias	
5.3.2 Changes in the Scale Factor That Depend on the Signal Amplitudes	
Summary	Q
References:	10
Appendix A – Calibration Signal Covariance Analysis	11
List of Tables	
Table 1. Measured Worst Case Harmonic Distortion	5
Table 2. Nonlinearity Coefficients Derived from Measured Harmonic Distortion	5
Table 3. Nonlinearity Coefficients Derived from the Required Harmonic Distortion	6
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# List of Acronyms, Abbreviations, and Symbols

$\beta_b$	London moment equivalent angle of		
	output bias		
$\beta_{dc}$	D.C. component of London moment		
	equivalent angle		
β <sub>c</sub>	Amplitude of London moment		
	equivalent angle at calibration signal		
	frequency		
β <sub>L</sub>	Amplitude of London moment		
	equivalent angle at satellite roll		
	frequency		
$\beta_L^{max}$	Full scale London moment equivalent		
	angle for a given range and gain of		
	SQUID readout electronics		
$\varepsilon_2, \varepsilon_3, \dots$	Nonlinearity coefficients		
δ[ ]	Change in quantity within brackets		
$\delta C_f$	Change in scale factor		
фо	Orbital Phase		
фа	Annual Phase		
$\Phi_{\rm in}$	Flux Input		
$\Phi_{ ext{max}}$	Maximum Input Flux		
σ	Standard error in each measurment		
$\sigma_{A,R,S,C}$	Standard error in A, R, S, and C		
ω	Angular Frequency		
$\omega_{\rm c}$	Angular calibration signal frequency		
$\omega_{\rm r}$	Angular satellite roll frequency		
Α	Amplitude		
$A_{c}$	Amplitude of calibration signal		
A <sub>o</sub>	Amplitude of orbital aberration		
$B_L$	London moment magnetic field		
$B_{max}$	Maximum trapped magnetic field for		
	a given range		
B <sub>T</sub>	Trapped magnetic field		
b	Voltage Bias		
C	Amplitude of calibration signal at		
	cosine of annual frequency		

Cr Scale Factor – flux in pickup loop to measured voltage output  EW Gyroscope misalignment perpendicular to plane of orbit (East-West)  EWo Gyroscope misalignment perpendicular to plane of orbit (East-West) at time to plane of orbit (North-South) number of measurements  NS Gyroscope misalignment in plane of orbit (North-South) at time to plane of orbit (North-South) at time to plane of orbit (North-South) at time to plane of orbit (East-West)  R Rate of Change of Calibration Signal Amplitude  Rew Gyroscope drift rate perpendicular to plane of orbit (East-West)  RNS Gyroscope drift rate in plane of orbit (North-South)  S Amplitude of calibration signal at sine of annual frequency  SQUID Superconducting Quantum Interference Device  t Time  ta Annual period  ti Time at ith sample  to Reference time  Vout Output Voltage  x states  z Measured SRE signal  zi Sampled SRE Signal at time ti				
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	t			
	ta	Annual period		
	t <sub>i</sub>	Time at i <sup>th</sup> sample		
<ul><li>x states</li><li>z Measured SRE signal</li></ul>		Reference time		
z Measured SRE signal	Vout	Output Voltage		
	X	states		
z <sub>i</sub> Sampled SRE Signal at time t <sub>i</sub>	Z			
	Zi	Sampled SRE Signal at time t <sub>i</sub>		

#### 1. Introduction

Requirement 5 of the Gravity Probe B Twelve Fundamental Requirements (T002) specifies the requirements on the noise and nonlinearity of the gyroscope readout system.

### Requirement 5, Gyroscope Readout

The SG readout system shall have a linearity consistent with an error of less than 0.3 marcsec/year. The gyroscope readout system single-sided noise spectral density shall be less than 190 marcsec/sqrt(Hz) x sqrt(roll period/180 sec) x (130 Hz/spin speed).

This paragraph contains two requirements. The first requirement states that the nonlinearity of the gyroscope readout system shall be consistent with an error of less than 0.3 marcsec/year The second requirement specifies the readout noise. One purpose of this document is to verify the requirement on the nonlinearity of the gyroscope readout.

A second purpose of this document is to justify the Gravity Probe B Program Control Board Change Request #559, which changed the requirements on the allowable harmonic distortion of in the System Design and Performance Requirements (T003). With this change these requirements read

#### 3.2.1 Linearity

The readout system shall meet the following linearity requirements in the presence of trapped flux levels as specified in requirement 1.5:

### 3.2.1.1 High Frequency

In the frequency range 100-1000 Hz the harmonic distortion of each of harmonics 2-5 of sinusoids with amplitude corresponding to the trapped flux levels of section 1.5 shall be less than 1.5e-4.

### 3.2.1.2 Low Frequency

For frequencies less than 1 Hz, the harmonic distortion of each harmonic 2-5 of sinusoids with amplitude up to 80 arcsec London moment equivalent shall be less than 1.5e-4.

For both requirements 3.2.1.1. and 3.2.1.2, the Program Control Board Change Request changed the maximum harmonic distortion from 1.0e-4 to 1.5e-4. This document demonstrates that T002, Requirement 5, is still satisfied with the increase of the allowable harmonic distortion.

### 2. Measurements of Harmonic Distortion

The nonlinearity of any system may be measured by injecting a sinusoidal signal into the input and measuring the harmonic distortion of the output. The System Design and Performance Requirements (T003) specify the harmonic distortion of a sinusoidal signal. These measurements are summarized in the "SRE Special Linearity Test Analysis", [1]. A low frequency signal at the satellite roll rate and a high frequency signal at the gyroscope spin frequency were injected into the pickup loop circuit and the harmonic distortion of the output was measured. The worst-case harmonic distortion while operating the SQUID Readout Electronics (SRE) in Range 1 or Range 2 is shown in Table 1 below. During science data collection, the SRE electronics will be operated on one of these two ranges. In the special linearity test, there was some

evidence of aliasing of high frequency signals which increased the measured harmonic distortion. The actual harmonic distortion is expected to be smaller than the values given here.

Table 1. Measured Worst Case Harmonic Distortion

	Worst Case Harmonic Distortion (PPM)			
	Harmonic 2	Harmonic 3	Harmonic 4	Harmonic 5
Low Frequency	135	102	104	79
High Frequency	116	91	44	39

# 3. Relationship of Nonlinearity to Harmonic Distortion

The nonlinearity of any system may be specified by a set of nonlinearity coefficients  $\epsilon_i$ , which are defined for a given maximum input. If the input is  $\Phi_{in}$  and the maximum input is  $\Phi_{max}$ , the output will be given by

$$V_{out} = C_f \Phi_{in} \left[ 1 + \varepsilon_2 \frac{\Phi_{in}}{\Phi_{max}} + \varepsilon_3 \left( \frac{\Phi_{in}}{\Phi_{max}} \right)^2 + \dots \right] + b$$
 (1)

where C<sub>f</sub> and b are respectively the scale factor and the bias of the readout. If the input is sinusoidal,

$$\Phi_{in} = A\cos\omega t$$

with amplitude A and angular frequency  $\omega$ , the output will be given by

$$V_{out} = C_f \begin{bmatrix} A \left( \frac{\varepsilon_2}{2} \frac{A}{\Phi_{\text{max}}} + \frac{3\varepsilon_4}{8} \frac{A^3}{\Phi_{\text{max}}^3} + \dots \right) + A\cos\omega t \left( 1 + \frac{3\varepsilon_3}{4} \frac{A^2}{\Phi_{\text{max}}^2} + \frac{5\varepsilon_5}{8} \frac{A^4}{\Phi_{\text{max}}^4} + \dots \right) + \\ + A\cos2\omega t \left( \frac{\varepsilon_2}{2} \frac{A}{\Phi_{\text{max}}} + \frac{\varepsilon_4}{2} \frac{A^3}{\Phi_{\text{max}}^3} + \dots \right) + A\cos3\omega t \left( \frac{\varepsilon_3}{4} \frac{A^2}{\Phi_{\text{max}}^2} + \frac{5\varepsilon_5}{16} \frac{A^4}{\Phi_{\text{max}}^4} + \dots \right) + \\ + A\cos4\omega t \left( \frac{\varepsilon_4}{8} \frac{A^3}{\Phi_{\text{max}}^3} + \dots \right) + A\cos5\omega t \left( \frac{\varepsilon_5}{16} \frac{A^4}{\Phi_{\text{max}}^4} + \dots \right) + \dots$$

From this equation, it can be seen that the measured harmonic distortion may be used to place limits on the nonlinearity coefficients. Since the amplitude of the input signal, A, is always less than  $\Phi_{max}$ , the first term in the expression for the amplitude of each harmonic will dominate as long as the nonlinearity coefficients are approximately the same size. With this approximation, the ratio of the amplitude at the  $n^{th}$  harmonic to the fundamental will be given by

$$\frac{A(n\omega)}{A(\omega)} \approx \frac{\varepsilon_n}{2^{n-1}} \frac{A^{n-1}}{\Phi_{\max}^{n-1}}$$

Then, the magnitude of any one of the nonlinearity coefficients may be estimated by

$$\varepsilon_n \approx 2^{n-1} \frac{\Phi_{\text{max}}^{n-1}}{A^{n-1}} \frac{A(n\omega)}{A(\omega)}$$

Using this equation, the worst-case nonlinearity coefficients may be determined from the measured harmonic distortion. For the high frequency signals, the sinusoidal signal was 90% of the maximum amplitude, while for the low frequency signal the maximum amplitude was equivalent to an angular misalignment of 80 arc seconds or approximately 80% of the full scale range. These worst-case nonlinearity coefficients are listed in Table 2. To verify requirement T002, #5, on the linearity of the gyroscope readout,

it is necessary to show that this requirement is met, using the values of the nonlinearity coefficients given in Table 2.

Table 2. Nonlinearity Coefficients Derived from Measured Harmonic Distortion

	$\epsilon_2$	ε3	ε <sub>4</sub>	E <sub>5</sub>
Low Frequency	$3.4 \times 10^{-4}$	$6.4 \times 10^{-4}$	$1.6 \times 10^{-3}$	$3.1 \times 10^{-3}$
High Frequency	$2.6 \times 10^{-4}$	$4.5 \times 10^{-4}$	$4.8 \times 10^{-4}$	$9.5 \times 10^{-4}$

For comparison the nonlinearity coefficients derived from the revised harmonic distortion requirements in System Design and Performance Requirements (T003) are listed in Table 3. Note that the difference between required values for the low frequency and high frequency distortion coefficients is because the amplitude of the injected signal is assumed to be 80% of full scale for the low frequency measurements and 90% of full scale for the high frequency measurements. To justify the change in T003 requirements 3.2.1.1 and 3.2.1.2, it is necessary to show that T002, Requirement 5 is met with the values of the nonlinearity coefficients given in Table 3.

Table 3. Nonlinearity Coefficients Derived from the Required Harmonic Distortion

	$\epsilon_2$	$\epsilon_3$	ε4	Es
Low Frequency	$3.75 \times 10^{-4}$	$9.4 \times 10^{-4}$	$2.3 \times 10^{-3}$	$5.9 \times 10^{-3}$
High Frequency	$3.3 \times 10^{-4}$	$7.4 \times 10^{-4}$	$1.6 \times 10^{-3}$	$3.7 \times 10^{-3}$

# 4. Effects of Nonlinearity on the Science Gyroscope Readout

Nonlinearity of the gyroscope readout will produce two distinctly different effects on the output. First, it will mix the various input signals to produce output signals at frequencies which are the sums and differences of the input signals. Secondly, it will change the scale factor of the system. The effect of the nonlinearity of the gyroscope readout system on the signals at the harmonics of the gyroscope spin speed are discussed in reference [2], and the effects of the nonlinearity of the gyroscope readout system on the signals at the gyroscope roll rate are discussed in reference [3]. The effects discussed in each of these documents is summarized here.

# 4.1 Effects of Nonlinearity on the Signals at the Harmonics of the Gyroscope Spin Speed

### 4.1.1 Rectification of High Frequency Signals

In [2], the London moment equivalent angle of the d.c. rectified signal due to the signals at the harmonics of the gyroscope spin speed was shown to be

$$\beta_{dc} \le \frac{\varepsilon_2}{2} \frac{B_T^2}{B_L B_{\text{max}}}$$

where  $\varepsilon_2$  is the quadratic high frequency nonlinearity coefficient given in Table 2,  $B_T$  is the trapped magnetic flux,  $B_L$  is the London moment field, and  $B_{max}$  is the trapped magnetic field which gives a full scale readout on a given range. Using the values  $B_T=8.3~\mu gauss$  (worst-case) [4],  $B_L=100~\mu gauss$  at 150 Hz spin speed [5],  $B_{max}=9~\mu gauss$  in range 2 (T003, Req. 1.5 and [6]), the London moment equivalent angle is 2.1 arc seconds. Since the magnitude of signals at the odd harmonics of the gyroscope spin speed will be modulated at the polhode frequency, this d.c.bias will also be modulated at the polhode frequency. Note that

this effect has a similar time signature as the effect of a misalignment of the plane of the pickup loop from the satellite roll axis [5]. Both of these effects produce a modulation of the output of the gyroscope readout system at harmonics of the polhode frequency.

The satellite roll rate will be chosen to be incommensurate with the polhode frequency. If these signals at the polhode frequency are not included in the data analysis, there is the potential for leakage from the harmonics of polhode frequency into the measured amplitude of the roll frequency signal [7]. This leakage depends on the amplitude of the interfering signal, the frequency difference between the signal of interest and the interfering signal, and the length of the data set. For sufficiently small amplitudes at frequencies well separated in comparison to the inverse of the length of the data set, the leakage will be small enough that it can be neglected. If there are harmonics of the polhode frequency close to the satellite roll frequency, and they are included in the data analysis, their effects will be effectively removed. In either case, there will be no contribution to the roll frequency amplitude or phase from this modulation.

# 4.1.2 Modulation of the Low Frequency Scale Factor

Also in [2], the modulation of the magnetic flux applied to the SQUID at the harmonics of the gyroscope spin speed and the nonlinearity of the SQUID can change the low frequency scale factor of the gyroscope readout. This effect will have no impact on the overall experiment error because the magnitude of the fractional change in the scale factor is less than 10<sup>-7</sup> and because the scale factor averaged over the polhode motion is expected to be constant.

# 4.2 Effects of Nonlinearity on Low Frequency Signals

### 4.2.1. Linearity of the Measurement Model

As shown in [3], the linear approximations made in the derivation of the measurement model introduce no additional errors greater than  $10^{-13}$  radians in the bias. These errors are small enough to be neglected.

# 4.2.2 Changes in the Scale Factor That Depend on the Bias

The changes in the scale factor of the low frequency gyroscope readout may be classified as those that depend on the bias and those that depend on the amplitude of the signals. The magnitude of the scale factor changes that depend on the bias are shown in [3] to be

$$\frac{\delta C_f}{C_f} = \varepsilon_2 \frac{2\beta_b}{\beta_L^{\text{max}}} + \varepsilon_3 \frac{3\beta_b^2}{\left(\beta_L^{\text{max}}\right)^2} + \dots$$

where  $\beta_b$  is the London moment equivalent angle of any bias in the output of the SQUID readout electronics and  $\beta_L^{max}$  is the full scale London moment equivalent angle for a given range and gain of the SQUID readout electronics. An on-board Science Mode Offset Adjust Algorithm [8] keeps the bias within a specified fraction of the maximum value and thereby reduces these changes in the scale factor. If the Offset Adjustment Algorithm keeps the bias within 5% of the full scale range and the nonlinearity coefficients derived from the required values for the harmonic distortion are used, the maximum change in the scale factor is less than  $4.4 \times 10^{-5}$ . Using the nonlinearity coefficients derived from the measured harmonic distortion, the maximum fractional change in the calibration signal is  $3.9 \times 10^{-5}$ . The linear drift in the scale factor and the annual variation in the scale factor will be a small fraction of this maximum change.

In addition, since the scale factor change at the calibration signal frequency is the same as the scale factor change at the satellite roll rate, the calibration signal may be used to monitor the change in the scale factor at the satellite roll frequency. Appendix A is a covariance analysis of the accuracy with which a linear drift in the scale factor and an annual variation in the scale factor may be determined from measurements of the calibration signal over the course of one year. With a SQUID noise of 190 mas/ $\sqrt{\text{Hz}}$  and a calibration signal with a London moment equivalent amplitude of 30 arc seconds, the amplitude of the calibration signal may be determined to a fractional accuracy of  $1.6 \times 10^{-6}$ . A linear drift and a sinusoidal variation may also be estimated at the same time. The statistical error in the amplitude of each of these variations divided by the amplitude of the calibration signal is respectively  $5.5 \times 10^{-6}$  and  $2.3 \times 10^{-6}$ .

The effects of an unmodeled linear drift or annual variation in the scale factor have been calculated in [9]. An unmodeled fractional linear drift in the scale factor of  $5.5 \times 10^{-6}$  will produce an error in the drift rate measurement of less than 0.07 mas/yr. An unmodeled sinusoidal variation in the scale factor of  $2.3 \times 10^{-6}$  will produce a error in the measurement of the drift rate of less than 0.04 mas/yr. Both of these errors depend on the start date of the data acquisition, and in each case the worst-case start date was used. Since these errors may be correlated, the combined error is taken to be the sum of these contributions, or 0.11 mas/yr. This value is a worst-case error associated with the bias variation and the nonlinearity.

There are several potential methods for further reducing this error in the data analysis. Since the bias is well known at any given time, it would be possible to look for a correlation between the bias and the scale factor. To the extent that this relation between the bias and scale factor are well known, the effects of bias variation on the scale factor may be removed. Another possibility is to search for a linear or annual variation in the scale factor in the data. Since the scale factor is multiplicative, these temporal variations produce additional time signatures that will be observable in the data.

# 4.2.3 Changes in the Scale Factor That Depend on the Signal Amplitudes

In addition to the changes in the scale factor that are proportional to the bias, the cubic nonlinearity terms produce a change in the gyroscope readout scale factor which depends on the amplitude of the roll frequency and calibration frequency signals themselves. As shown in [3], the fractional change in the scale factor at the satellite roll frequency will be

$$\frac{\delta C_f(\omega_r)}{C_f} = \varepsilon_3 \, \delta \left[ \frac{3}{4} \left( \frac{\beta_L}{\beta_L^{\text{max}}} \right)^2 + \frac{3}{2} \left( \frac{\beta_C}{\beta_L^{\text{max}}} \right)^2 \right] + \dots$$

and the fractional change in the scale factor at the calibration signal frequency is given by

$$\frac{\delta C_f(\omega_C)}{C_f} = \varepsilon_3 \, \delta \left[ \frac{3}{2} \left( \frac{\beta_L}{\beta_L^{\text{max}}} \right)^2 + \frac{3}{4} \left( \frac{\beta_C}{\beta_L^{\text{max}}} \right)^2 \right] + \dots$$

where  $\delta[\ ]$  denotes the change in the quantity within the brackets. Since the change in the scale factor at the satellite roll frequency is not the same as the change in the scale factor at the calibration signal frequency, the amplitude of the calibration signal may not be used to monitor changes in the scale factor at the satellite roll frequency.

The magnitude of this contribution to the change in the scale factor may be estimated from the expected changes in the gyroscope spin axis. These expected changes include the effects of an average misalignment,  $NS_0$  and  $EW_0$ , the linear drift rate of the gyroscope,  $R_{NS}$  and  $R_{EW}$ , and the orbital and annual

aberration signals. The effects of the gravitational deflection of light and parallax are small enough that they may be neglected here. Then,

$$\begin{split} NS &= NS_o + R_{NS}(t-t_0) + A_o \cos\phi_o + l_1 \cos\phi_a + l_2 \sin\phi_a \\ EW &= EW_o + R_{EW}(t-t_o) + l_3 \cos\phi_a + l_4 \sin\phi_a \end{split}$$

Here,  $A_0$  is the amplitude of the orbital aberration signal, and  $l_1$ ,  $l_2$ ,  $l_3$ , and  $l_4$  are the components of the annual aberration signal. The phase of the orbital motion is denoted by  $\phi_0$  and the annual motion by  $\phi_a$ . When these terms are squared, only those changes in the scale factor at the orbital or annual frequencies or those that are linear in time will have a significant effect on the data analysis. The largest of these effects will be the term which arises from the average misalignment and the annual aberration. If the average misalignment is 1 arc sec (T002, Req. 6) and the amplitude of the annual aberration is 20 arc sec due to the motion of the earth about the sun, then the magnitude of the change in the scale factor at the annual frequency, will be given by

$$\frac{\delta C_f(\omega_r)}{C_f}\bigg|_{annual} = \varepsilon_3 \frac{3}{4} \frac{2\langle \beta_L \rangle_{average} (\beta_L)_{annual}}{(\beta_L^{\text{max}})^2} \approx 9.4 \times 10^{-4} \frac{3}{4} \frac{2(1 \text{ arc sec})(20 \text{ arc sec})}{(100 \text{ arc sec})^2} = 2.8 \times 10^{-6}$$

Using the results of [9], an unmodeled annual variation in the scale factor of this magnitude will give a worst case error in the gyroscope drift rate of 0.06 mas/yr. The value of nonlinearity coefficient used here is derived from the required value of the harmonic distortion. The measured values will further reduce this worst-case error.

As in the case of the scale factor variations that depend on the bias, there are additional possibilities for reducing these errors in the data analysis. The amplitude of the roll frequency signal will be well known at all times, and it will be possible to look for correlations between this amplitude and the scale factor. In addition, the data itself may be used to place limits on linear and annual scale factor variations.

### **Summary**

The effects of nonlinearity on the gyroscope readout system may be summarized as follows:

- 1. High frequency nonlinearity of the gyroscope readout system at the odd harmonics of the gyroscope spin frequency will mix the large signals from the trapped magnetic flux to produced signals at d.c. and the even harmonics of the gyroscope spin speed. The worst case amplitude of these signals is expected to be on the order of 2.1 arc sec. Since these signals are incommensurate with the satellite roll rate, they are expected to have a negligible effect on the measurement of the gyroscope drift rate.
- 2. The flux applied to the SQUID at the odd harmonics of the gyroscope spin speed and the nonlinearity of the SQUID itself will change the open loop gain of the flux locked loop. The change in the scale factor of the gyroscope readout system associated with this change in the open loop gain is too small to have any significant impact on the experiment error. In addition, the value of this altered scale factor averaged over the polhode motion is expected to be stable.
- 3. The linear approximations used to derive the model for the gyroscope motion are sufficiently accurate that higher order terms do not need to be included.

- 4. The changes in the scale factor that depend on the bias may be monitored using the calibration signal. The accuracy with which the changes in the calibration signal may be determined in the presence of noise will determine the residual unmodeled changes in the scale factor. These unmodeled changes in the scale factor, in turn, contribute to the unmodeled error in the gyroscope drift rate. The RSS of the errors due to the residual uncertainty in the linear and annual variations of the scale factor is less than 0.08 mas/yr. Note that this error is applies to both the required and measured values of the nonlinearity coefficients.
- 5. Changes in the scale factor which are proportional to the amplitude of the roll frequency and calibration frequency signal amplitudes will produce errors in the drift rate of less than 0.06 mas/yr as long as the T003 requirements  $(1.5 \times 10^{-4})$  on the harmonic distortion are met.

The magnitude of these error estimates are worst-case values. In addition, the effects are uncorrelated, so it is appropriate to take the root square sum (RSS) of these errors. The RSS of the error in the gyroscope drift rate due to the nonlinearity determined from the required value of the harmonic distortion  $(1.5 \times 10^{-4})$  is 0.10 mas/year. The corresponding error due to the nonlinearity determined from the measured values of the harmonic distortion is slightly less than this value. Several approaches, which are described above, may be used to reduce this error in the ground-based data analysis with the flight data.

#### References:

- 1. Lockhart, J.M., SRE Special Linearity Test Analysis, S0654, Rev. A, GP-B, Hansen Laboratories, Stanford University, 2003.
- 2. Keiser, G.M., Effects of High Frequency Nonlinearity on the Gyroscope Readout System, S0894, Rev. -, GP-B, Hansen Laboratories, Stanford University, 2003.
- 3. Mandel, I. and A.S. Silbergleit, On the Linearity of the Gravity Probe B Measurement Model, S0632, Rev. A, Hansen Laboratories, GP-B, Stanford University, 2003.
- 4. Whelan, R. and B. Muhlfelder, *Verification of Trapped Flux Requirements*, S0564 Rev. A, GP-B, Hansen Laboratories, Stanford University, 2001.
- 5. Keiser, G.M., J. Lockhart, and B. Muhlfelder, Bias Variations in the Gravity Probe B Readout System, S0352, Rev. -, GP-B, Hansen Laboratories, Stanford University, 2003.
- 6. Scarlett, S., Gravity Probe B Relativity Mission, SQUID Readout Electronics (SRE) Specification, LMMS/P480136D, Lockheed-Martin, 2002.
- 7. Haupt, G.T., Development and Experimental Verification of a Nonlinear Data Reduction Algorithm for the Gravity Probe B Relativity Mission, Ph. D., Stanford University, 1996.
- 8. Lockheed-Martin, Gravity Probe B Relativity Mission, Flight Software Design Specification, LMMS/P479922B, Lockheed-Martin, 2000.
- 9. Keiser, G.M., Analytical Solution for Unmodeled Errors in the Gravity Probe B Experiment, S0895, Rev. -, GP-B, Hansen Laboratories, Stanford University, 2003.
- 10. Keiser, G.M., Analytical Solution to the Gravity Probe B Covariance Matrix, S0351, Gravity Probe B, Stanford University, 1998.

## Appendix A - Calibration Signal Covariance Analysis

The calibration signal is injected into the pickup loop circuit and its amplitude and phase, as measured by the SQUID Readout Electronics, may be used to monitor any changes in the scale factor and phase shift of the gyroscope readout system at the calibration signal frequency. The purpose of this covariance analysis is to calculate with what accuracy the amplitude, a linear drift in the amplitude, and an annual sinusoidal variation in the amplitude of the measured calibration signal frequency may be estimated. The accuracy of these components of the change in the calibration signal may be used to estimate the accuracy with which the changes in the scale factor may be estimated. The methods used here are similar to those used to calculate the analytical solution of the Gravity Probe B covariance matrix [10].

If there is a linear drift or an annual sinusoidal variation in the measured amplitude at the calibration signal frequency, then the measured calibration signal may be modeled as

$$z = \left(A_c + R\frac{t}{t_a} + S\sin\frac{2\pi t}{t_a} + C\cos\frac{2\pi t}{t_a}\right)\cos\omega_c t$$

Here t is time,  $t_a$  is one year,  $\omega_c$  is the angular calibration signal frequency, and  $A_c$ , R, S, and C represent the amplitude, the rate of change of the amplitude, and variation in the amplitude and the sine and cosine of the annual frequency. The relation between any given measurement,  $z_i$ , these four states may be written as

$$z_i = H_i x$$

where H<sub>i</sub> is the given by

$$H_i = \begin{bmatrix} 1 & t_i/t_a & \sin(2\pi t_i/t_a) & \cos(2\pi t_i/t_a) \end{bmatrix}$$

and x is  $4 \times 1$  matrix of the states of the system.

$$x^T = [A_c R S C]$$

The covariance matrix of the four states will then be given by

$$P = \sum_{i} \frac{H_i^T H_i}{\sigma^2}$$

where  $\sigma$  is the standard error in each measurement of z. This covariance matrix may be calculated by converting the sum to an integral and integrating over one year. For a one year period the off-diagonal terms are found to be zero. The diagonal terms give the standard error for each of the four states:

$$\sigma_{A_c}^2 = 2\sigma^2 / N$$

$$\sigma_R^2 = 24\sigma^2 / N$$

$$\sigma_S^2 = \sigma_C^2 = 4\sigma^2 / N$$

where N is the number of measurements of z.

The standard error of each of these parameters may be estimated for the data taken over one year assuming the noise in the gyroscope readout system only meets the required value of  $190 \text{ mas}/\sqrt{\text{Hz}}$ . If the measurements are made at one second intervals over the course of one year,  $\sigma$ =190 mas and N=3.15 × 10<sup>7</sup>. The data for both guide star valid and guide star invalid may be used since the calibration signal may be used during both those periods. If the calibration signal has an amplitude equivalent to a London moment signal of 30 arc sec, then the corresponding fractional changes is the scale factor are

$$\left(\frac{\delta C_f}{C_f}\right)_R = \frac{\sigma_R}{A_c} = 5.5 \times 10^{-6}$$

$$\left(\frac{\delta C_f}{C_f}\right)_{S,C} = \frac{\sigma_{S,C}}{A_c} = 2.3 \times 10^{-6}$$

These values are used as the residual unmodeled linear and sinusoidal drift in the scale factor at an annual period.